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Distortion of a Magnetic Field by the Motion of a Cylindrical Conductor

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Abstract

Motion of a conductor relative to a magnetic field distorts the field. This paper considers a cylindrical slug moving in a two-dimensional magnetic field, represented by the vector potential $A_0 \hat{\theta}$, where $\partial A_0 / \partial \theta = 0$. Maxwell's equations are solved for the distorted potential $A \hat{\theta}$ in the form of a rapidly converging series $A = \sum A_n$. The A_n 's are given in a form suitable for evaluation by a digital computer. The nonequivalence of the apparently analogous problem of a stationary slug in a time-varying field is noted and discussed.

Contents

1. INTRODUCTION	1
1.1 Background	1
1.2 Problem	1
1.3 Method of Solution	3
2. SOLUTION	3
2.1 Mathematical Expression of Problem	3
2.2 Functions to be Used	6
2.3 Solution	10
3. CONCLUSIONS	13
ACKNOWLEDGMENTS	15
REFERENCES	15
APPENDIX A	A1

Illustrations

1. Geometry of Problem	2
2. A Plot of the Function $f(x_n) = x_n^{-1/2} [(2 - x_n) K_n - 2 E_n]$ vs. x_n	8

3. Curves which Map the $\hat{\theta}$ Component of the Vector Potential $a_0(x_0)$
Due to Unit Current Flowing in a Loop of Unit Radius 9
4. $A_1(2\pi/\alpha)$, $A_2(2\pi/\alpha)^2$, and $A_3(2\pi/\alpha)^3$ as Functions of Slug Position
for a Slug of Radius $b = 0.67$, and Length $l = 3.33$ 14

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1. INTRODUCTION

1.1 Background

By Faraday's law a change of the magnetic field in a conducting slug induces currents in that slug. This change may be either time variation of the field on a stationary slug or motion of the slug in a constant field. By Ampere's law these currents produce an induced magnetic field which perturbs the original field. Such distortions have been used to measure the conductivity of various moving (Lin, Resler and Kantrowitz, 1955) and stationary (Chambers and Park, 1961) slugs. Theoretical analysis has been limited to a rectangular slug in a one-dimensional field (Oddson, 1963). These devices have been calibrated by measuring slugs of known conductivity. However, some phenomena, such as skin effect, which were not a problem in these cases, can be difficult to duplicate in the calibrating slugs. To extrapolate the calibration to these cases, one must solve the problem for a cylindrical slug in a two-dimensional field.

1.2 Problem

Consider an initial field represented by the vector potential $A_0 \hat{\theta}$ with the condition that $\partial A_0 / \partial \theta = 0$. A cylindrical slug of radius b , length 2ℓ , and

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conductivity σ , moves into this field at velocity v . (Figure 1.) The slug's midpoint is at position $z = vt$ at time t . What perturbation in A will be caused as a result of this motion? Since this problem involves spatial variations of the field, it will be referred to as the Space Case or Space. Consider also the problem of an identical but stationary slug with its midpoint at the point $z_0 = \text{const.}$ in an initial field represented by the vector potential $A_0(t) \hat{\theta}$. Again with the condition that $\partial A_0 / \partial \theta = 0$. What perturbation in A will be caused as a result of this time variation of the initial field? Since this problem involves temporal variations of the field, it will be referred to as the Time Case or Time.

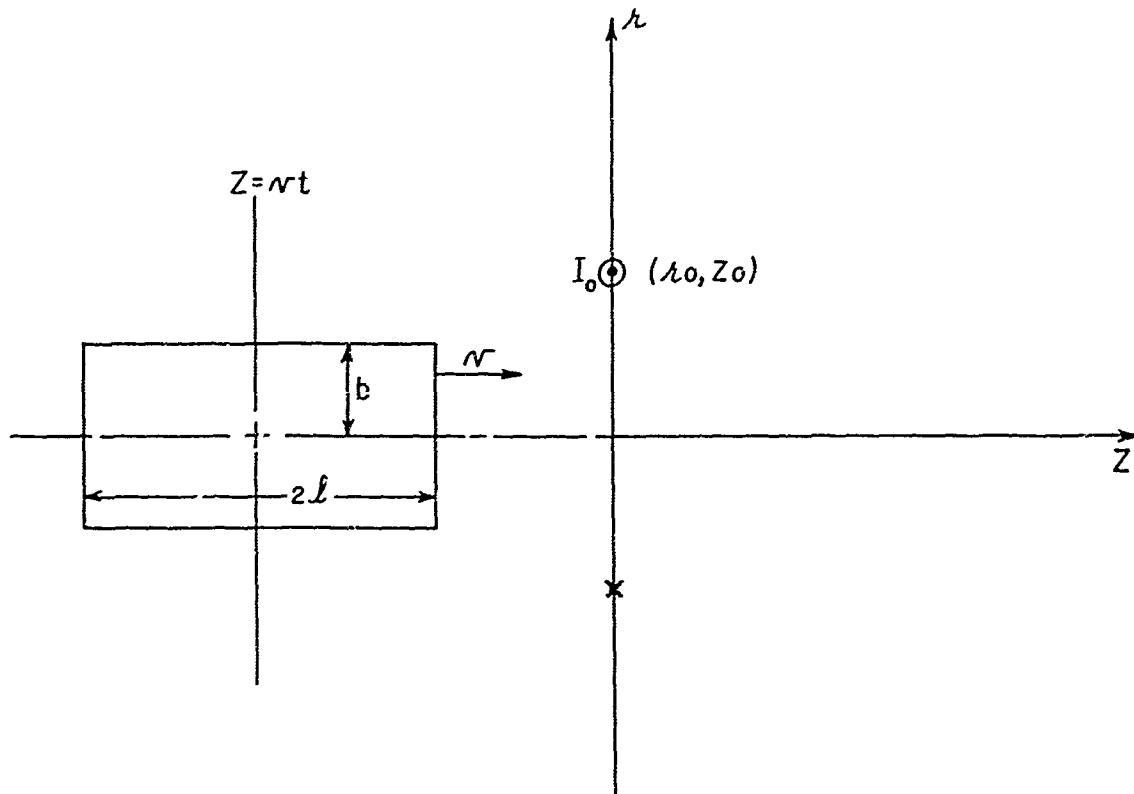


Figure 1. Geometry of Problem. A cylindrical slug of Length $2l$ and radius b moves along the z axis with velocity v . At time t the midpoint of the slug is at the point where $z = vt$. A current I_0 flows in a single turn loop at $r_0, z_0 = 0$

For the purpose of this discussion let A_0 be created by current I_0 flowing in the loop at $r_0, z_0 = 0$. More complex fields of this symmetry may be represented by a number of such loops.

1.3 Method of Solution

The symmetries postulated above allow one to reduce Maxwell's equations for moving media to one scalar differential equation for A , where A is the vector potential for the total magnetic field, both initial and induced. Let A equal $\sum A_n$ where A_n is the n^{th} order perturbation. Then, by physical reasoning, A_0 generates first order eddy currents I_1 which create the first order perturbing field A_1 . Similarly potential A_1 generates currents I_2 which create potential A_2 , and so on. This can be carried to as many orders as accuracy requires. Rationalized mks units will be used initially. Later all quantities will be made dimensionless.

2. SOLUTION

2.1 Mathematical Expression of Problem

Consider Maxwell's equations for moving media (Panofsky and Phillips, 1955).

$$\nabla \cdot \underline{E} = 0$$

$$\nabla \cdot \underline{B} = 0$$

$$\nabla \times \underline{B} = \mu_0 \underline{j}_0 + \mu_0 \sigma \underline{E} + \mu_0 \sigma \underline{v} \times \underline{B} + \mu_0 \epsilon_0 (\partial \underline{E} / \partial t) \quad (\text{Space})$$

$$\nabla \times \underline{E} = -\mu_0 \epsilon_0 (\partial \underline{B} / \partial t) \quad (\text{Time})$$

$$\nabla \times \underline{E} = -\partial \underline{B} / \partial t \quad (1)$$

\underline{j}_0 is the current density which produces the initial field.

σ is the conductivity of the medium.

\underline{v} is the velocity of the medium relative to the fields.

Charges, both true and those due to polarization, are assumed to be nonexistent. The velocity of the medium is much less than the velocity of light. Maxwell's equations outside of the medium are obtained, if the velocity and conductivity are set equal to zero.

The fields may be expressed by the vector potential \underline{A} .

$$\underline{E} = -\partial \underline{A} / \partial t$$

$$\underline{B} = \nabla \times \underline{A} \quad (2)$$

Combining Eq. (2) and Eq. (1), one finds that

$$\nabla \times (\nabla \times \underline{A}) = \mu_0 \underline{j}_0 - \mu_0 \sigma (\partial \underline{A} / \partial t) + \mu_0 \sigma \underline{v} \times (\nabla \times \underline{A}) - \mu_0 \epsilon_0 (\partial^2 \underline{A} / \partial t^2) \quad (\text{Space})$$

$$\nabla \times (\nabla \times \underline{A}) = \mu_0 \underline{j}_0 - \mu_0 \sigma (\partial \underline{A} / \partial t) - \mu_0 \epsilon_0 (\partial^2 \underline{A} / \partial t^2). \quad (\text{Time}) \quad (3)$$

For this problem, the vectors have the following forms:

$$\underline{j}_0 = j_0 \hat{\theta} \quad (4a)$$

$$\underline{v} = v \hat{k} \quad (4b)$$

$$\underline{B} = B_r \hat{r} + B_z \hat{k} \quad (4c)$$

$$\underline{E} = E \hat{\theta} \quad (4d)$$

due to the symmetries postulated above.

From Eqs. (2), (4c) and (4d), it follows that

$$\underline{A} = A \hat{\theta} \quad (5)$$

since it has been assumed that

$$\partial \underline{E} / \partial \theta = \partial \underline{B} / \partial \theta = 0 \quad (6)$$

then

$$\partial \underline{A} / \partial \theta = 0. \quad (7)$$

From these considerations it can be shown that only the $\hat{\theta}$ component of Eq. (3) is important.

$$\begin{aligned} (\partial / \partial r) \left[r^{-1} (\partial / \partial r) (r A) \right] + (\partial^2 A / \partial z^2) = -\mu_0 j_0 + \mu_0 \sigma v (\partial A / \partial z) \\ + \mu_0 \sigma (\partial A / \partial t) + \mu_0 \epsilon_0 (\partial^2 A / \partial t^2) \end{aligned} \quad (\text{Space})$$

$$\begin{aligned} (\partial / \partial r) \left[r^{-1} (\partial / \partial r) (r A) \right] + (\partial^2 A / \partial z^2) = -\mu_0 j_0 + \mu_0 \sigma (\partial A / \partial t) \\ + \mu_0 \epsilon_0 (\partial^2 A / \partial t^2) \end{aligned} \quad (\text{Time}) \quad (8)$$

Note that $\mu_0 \epsilon_0 = c^{-2}$, where c is the velocity of light. The term $(\partial^2 A / \partial z^2)$ is of the same order as $v^{-2} (\partial^2 A / \partial t^2)$ (Space) or $\omega_0^{-2} r_0^{-2} (\partial^2 A / \partial t^2)$ (Time). The frequency ω_0 is some fundamental frequency of the Time Case so chosen that $\omega_0 r_0$ (Time) is on the order of v (Space). Thus the term $\mu_0 \epsilon_0 (\partial^2 A / \partial t^2) - (v/c)^2 v^{-2} (\partial^2 A / \partial t^2)$ (Space) = $(\omega_0 r_0 / c)^2 \omega_0^{-2} r_0^{-2} (\partial^2 A / \partial t^2)$ (Time) may be omitted.

This makes Eq. (8) a diffusion equation. Since physically this is a diffusion problem, the approximation is valid.

All quantities will now be made dimensionless as follows: Choose a characteristic current I_0 and a characteristic length r_0 . Multiply each term in Eq. (8) by $r_0^2 (\mu_0 I_0)^{-1}$. Substitute dimensionless terms for the resulting ratios as indicated in Table 1. For the remainder of this paper only these dimensionless terms will be used. The dimensionless equivalent of Eq. (8) is

$$\begin{aligned} (\partial / \partial r) [r^{-1} (\partial / \partial r) (r A)] + (\partial^2 A / \partial z^2) \\ = -j_0 + \alpha [(\partial A / \partial z) + (\partial A / \partial t)] \quad (\text{Space}) \\ = -j_0 + \alpha [\partial A / \partial t]. \quad (\text{Time}) \quad (9) \end{aligned}$$

Table 1. Relationship Between the Ratios of Quantities in mks Units and the New Dimensionless Quantities

Ratios	Dimensionless Quantities
r/r_0	r
z/r_0	z
vt/r_0 (Space)	t
$\omega_0 t$ (Time)	
$A (\mu_0 I_0)^{-1}$	A
$j_0 r_0^2 / I_0$	j_0
b/r_0	b
ℓ/r_0	ℓ
r_0/r_0	1

Table 1. Relationship Between the Ratios of Quantities in mks Units and the New Dimensionless Quantities (Cont)

Ratios	Dimensionless Quantities
z_0/r_0	z_0
$\mu_0 \sigma v r_0$ (Space)	α
$\mu_0 \sigma \omega_0 r_0^2$ (Time)	

2.2 Functions to be Used

Consider first the potential a_n due to a current I_n in a loop r_n, z_n . The current density j_n is given by

$$j_n = I_n \delta(z - z_n) \delta(r - r_n). \quad (10)$$

The Dirac delta, $\delta(y - y_m)$, is defined by Eq. (11).

$$\begin{aligned} \int_{y_1}^{y_2} \delta(y - y_m) dy &= 1 & y_1 \leq y_m \leq y_2 \\ &= 0 & \text{otherwise} \end{aligned}$$

$$\begin{aligned} \int_{y_1}^{y_2} g(y) \delta(y - y_m) dy &= g(y_m) & y_1 \leq y_m \leq y_2 \\ &= 0 & \text{otherwise.} \end{aligned} \quad (11)$$

The potential is given by (Stratton, 1941)

$$a_n = (2\pi)^{-1} I_n (r_n/r)^{1/2} f(x_n). \quad (12)$$

The following definitions are required:

$$\begin{aligned} x_n &= 4 r_n r [(r_n + r)^2 + (z - z_n)^2]^{-1} \\ f(x_n) &= x_n^{-1/2} [(2 - x_n) K_n - 2 E_n] \\ K_n &= \int_0^{\pi/2} [1 - x_n \sin^2 \phi]^{-1/2} d\phi \\ E_n &= \int_0^{\pi/2} [1 - x_n \sin^2 \phi]^{1/2} d\phi \end{aligned} \quad (13)$$

K_n and E_n are the complete elliptic integrals of the first and second kinds (Jahnke and Emde, 1945). The function $f(x_n)$ is shown in Figure 2; $a_0(r_0, z_0)$ is mapped in Figure 3.

Azimuthal currents in a solid may be regarded as a continuous distribution of such rings. To obtain the potential due to such currents, one must integrate over all such rings and divide by an appropriate normalizing factor. Thus,

$$A_n(r, z) = \frac{\int_0^b \int_{t-\ell}^{t+\ell} a_n dz_n dr_n}{z b \ell} \quad (\text{Space})$$

$$= \frac{\int_0^b \int_{z_0-\ell}^{z_0+\ell} a_n dz_n dr_n}{z b \ell} \quad (\text{Time}) \quad (14)$$

It should be noted that

$$I_n = I_n(r_n, z_n). \quad (15)$$

Since the normalizing factor will later cancel, large values of ℓ present no problem. Note that n refers to the n^{th} -order perturbation. The rings are not discrete and therefore are not numbered

One more function is needed. We define $A_{n, n-1}$ to be the function A_{n-1} with r_n, z_n substituted for r, z . Similar definitions may be formed for $x_{n, n-1}, K_{n, n-1}$, and $E_{n, n-1}$. $a_{n, n-1}$ is the potential at r_n, z_n due to a current I_{n-1} in the loop at r_{n-1}, z_{n-1} .

Since a_n is the potential of a current loop, it must satisfy an equation similar to Eq. (9) except $\alpha = 0$.

$$(\partial/\partial r) [r^{-1} (\partial/\partial r) (r a_n)] + (\partial^2 a_n / \partial z^2) = -I_n \delta(r - r_n) \delta(z - z_n). \quad (16)$$

Since a differential with respect to r is unaffected by integration over r_n , Eqs. (14) and (16) show that

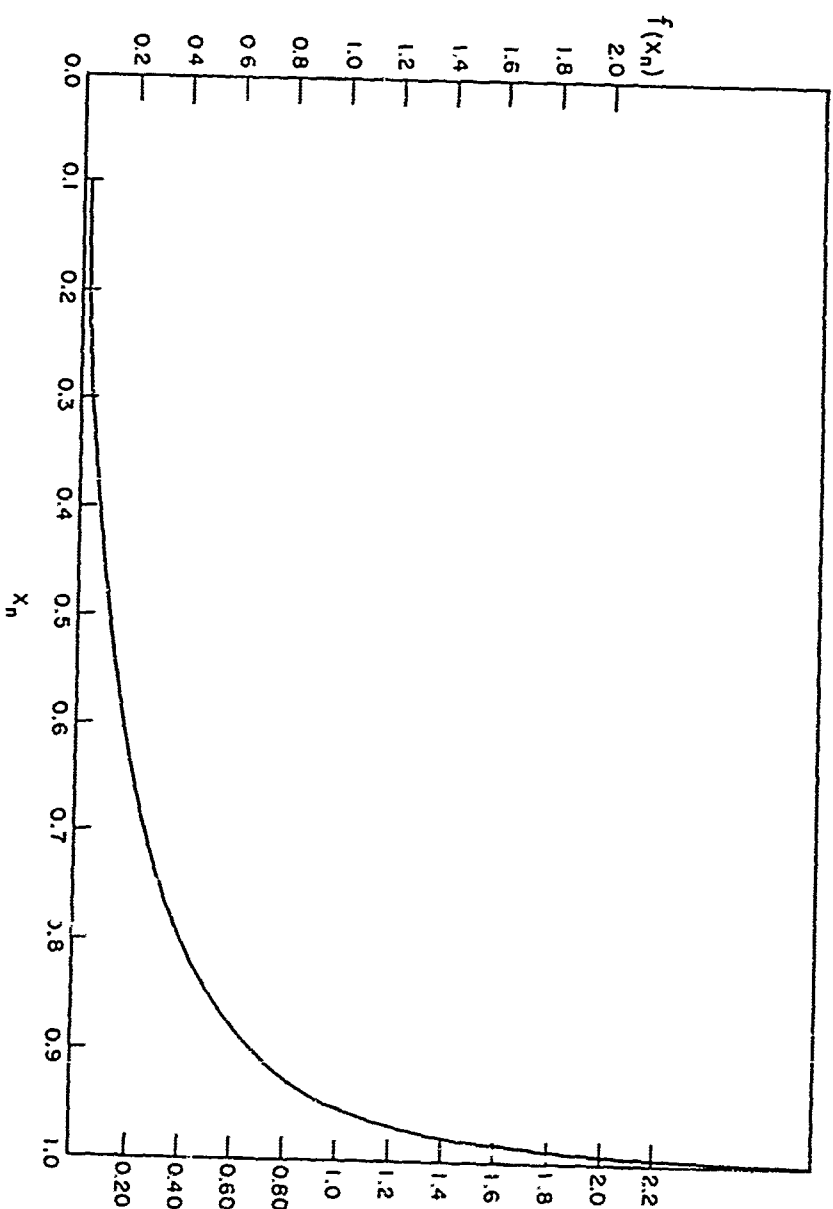


Figure 2. A Plot of the Function $f(x_n) = x_n^{-1/2} [(2 - x_n) K_n - 2 E_n]$ vs. x_n . The complete elliptic integrals K_n and E_n are defined: $K_n = \int_0^{\pi/2} [1 - x_n \sin^2 \phi]^{-1/2} d\phi$ $E_n = \int_0^{\pi/2} [1 - x_n \sin^2 \phi]^{1/2} d\phi$

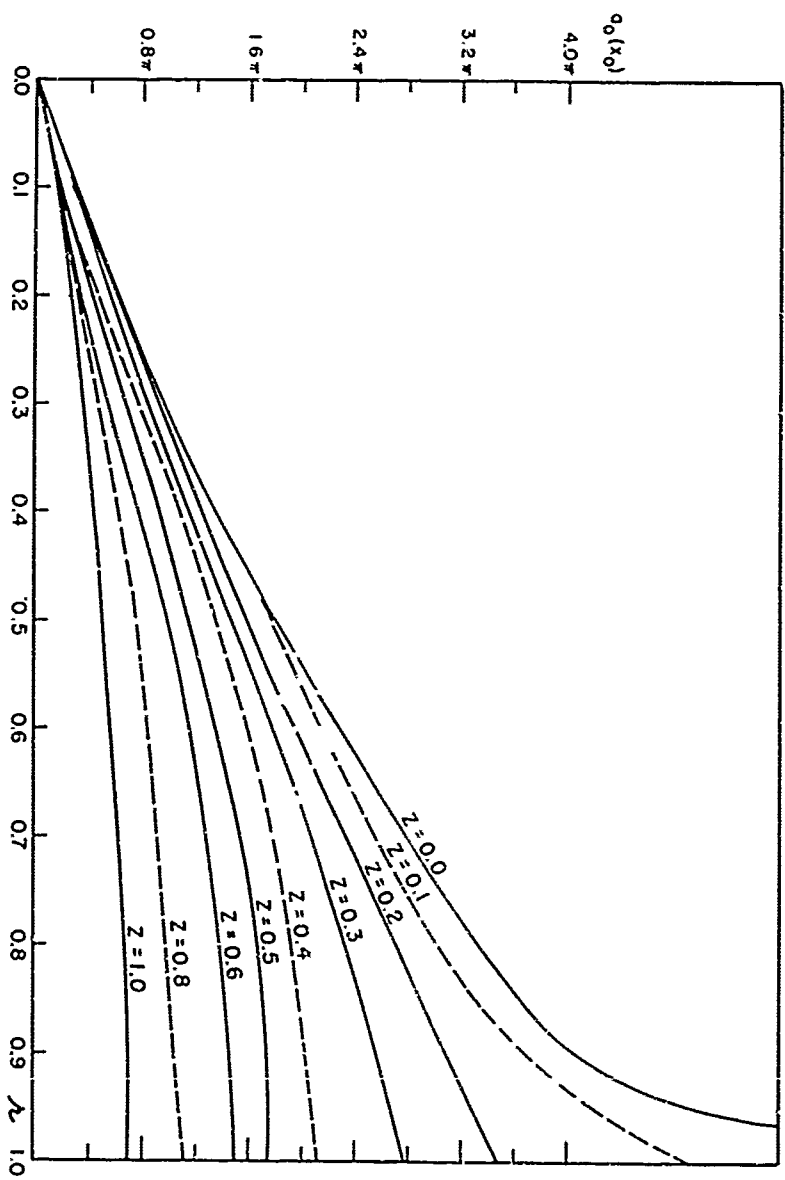


Figure 3. Curves which Map the $\hat{\theta}$ Component of the Vector Potential $a_{\theta}(x_0)$ Due to Unit Current Flowing in a Loop of Unit Radius. The potential is mapped as a function of distance r from the axis at various points z along the axis

$$\begin{aligned}
& (\partial/\partial r) [r^{-1} (\partial/\partial r) (r A_n)] + (\partial^2 A_n / \partial z^2) \\
& = -(2b\ell)^{-1} \int_0^b \int_{t-\ell}^{t+\ell} I_n(r_n, z_n) \delta(r - r_n) \delta(z - z_n) dz_n dr_n \quad (\text{Space}) \\
& = -(2b\ell)^{-1} \int_0^b \int_{z_0-\ell}^{z_0+\ell} I_n(r_n, z_n) \delta(r - r_n) \delta(z - z_n) dz_n dr_n \quad (\text{Time}) \quad (17)
\end{aligned}$$

Using the definition of $\delta(y - y_m)$ in Eq. (11), one may show that

$$\begin{aligned}
& (\partial/\partial r) [r^{-1} (\partial/\partial r) (r A_n)] + (\partial^2 A_n / \partial z^2) \\
& = -(2b\ell)^{-1} I_n(r, z) \quad r \leq b, \quad t - \ell \leq z \leq t + \ell \quad (\text{Space}) \\
& \quad \quad \quad z_0 - \ell \leq z \leq z_0 + \ell \quad (\text{Time}) \\
& = 0 \quad \text{otherwise.} \quad (18)
\end{aligned}$$

2.3 Solution

Let the solution to Eq. (9) be

$$A = \sum A_n. \quad (19)$$

The A_n 's will now be evaluated.

Outside of the slug, Eq. (9) becomes

$$(\partial/\partial r) [r^{-1} (\partial/\partial r) (r A)] + (\partial^2 A / \partial z^2) = -j_0. \quad (20)$$

If Eq. (19) is used to substitute for A on the left-hand side of Eq. (20), the A_0 term will contribute $-j_0$ to the right-hand side and the other A_n 's will contribute nothing to the right-hand side and the solution is obviously valid. Inside the slug Eq. (9) can be satisfied in the same manner if

$$\alpha [(\partial A / \partial z) + (\partial A / \partial t)] = -(2b\ell)^{-1} \sum_{n=1}^{\infty} I_n(r, z). \quad (21)$$

If Eq. (19) is now used to replace A on the left-hand side of Eq. (21)

$$\begin{aligned}
 (-2b\ell)^{-1} \sum_{n=1}^{\infty} I_n(r, z) &= \\
 \alpha \sum_{n=1}^{\infty} [(\partial A_{n-1}/\partial z) + (\partial A_{n-1}/\partial t)] &\quad \text{(Space)} \\
 \alpha \sum_{n=1}^{\infty} [\partial A_{n-1}/\partial t] &\quad \text{(Time) (22)}
 \end{aligned}$$

Replacing r, z by r_n, z_n and equating these two series term by term

$$\begin{aligned}
 I_n(r_n, z_n) &= -2\alpha b\ell [(\partial A_{n, n-1}/\partial z_n) + (\partial A_{n, n-1}/\partial t)] \quad \text{(Space)} \\
 &= -2\alpha b\ell [\partial A_{n, n-1}/\partial t] \quad \text{(Time) (23)}
 \end{aligned}$$

When this is substituted into Eq. (12) and the result substituted into Eq. (14)

$$\begin{aligned}
 A_n &= -(2\pi)^{-1} \alpha \int_0^b \int_{t-\ell}^{t+\ell} (r_n/r)^{1/2} [(\partial A_{n, n-1}/\partial z_n) + (\partial A_{n, n-1}/\partial t)] \\
 &\quad f(x_n) dz_n dr_n \quad \text{(Space)} \\
 &= -(2\pi)^{-1} \alpha \int_0^b \int_{z_0-\ell}^{z_0+\ell} (r_n/r)^{1/2} [\partial A_{n, n-1}/\partial t] f(x_n) dz_n dr_n \quad \text{(Time) (24)}
 \end{aligned}$$

An important difference in the two cases may now be noted. In the Time Case only $A_{n, n-1}$ is time dependent. Equation (24) may be simplified by taking the time derivative outside the integral.

$$A_n = (\partial/\partial t) (-2\pi)^{-1} \alpha \int_0^b \int_{z_0-\ell}^{z_0+\ell} A_{n, n-1} f(x_n) dz_n dr_n \quad \text{(Time) (25t)}$$

An iteration process will now yield

$$A_n = (\partial^n / \partial t^n) (-2\pi)^{-n} \alpha^n \int_0^b \int_0^b \dots \int_0^b \int_{z_0-\ell}^{z_0+\ell} \dots \int_{z_0-\ell}^{z_0+\ell} f(x_1) f(x_2) \dots f(x_n) dz_1 \dots dz_n dr_1 dr_2 \dots dr_n \quad (26t)$$

Usually the time dependence of the initial field can be separated $A_{1,0} = A'_{1,0}(r,z) g_{1,0}(t)$ to yield

$$A_n = (\partial^n g / \partial t^n) (-2\pi)^{-n} \alpha^n \int_0^b \int_0^b \dots \int_0^b \int_{z_0-\ell}^{z_0+\ell} \dots \int_{z_0-\ell}^{z_0+\ell} f(x_1) \dots f(x_n) dz_1 \dots dz_n dr_1 \dots dr_n \quad (27t)$$

If $g_{1,0}(t) = e^{it}$ then

$$A_n = (-i/2\pi)^n \alpha^n e^{it} \int_0^b \int_0^b \dots \int_0^b \int_{z_0-\ell}^{z_0+\ell} \dots \int_{z_0-\ell}^{z_0+\ell} f(x_1) \dots f(x_n) dz_1 \dots dz_n dr_1 \dots dr_n \quad (28t)$$

Thus each A_n is $\cos t$ or $\sin t$ times an amplitude.

In the Space Case such a simplification is not possible. There are z and t dependent terms in the integral and in the limits which are not to be differentiated. As will be seen below, the successive A_n differ by more than simple phase shifts. However, the differential equation (Eq. (9)) has been changed to an infinite number of integral equations (Eqs. (19) and (24)). The n^{th} equation corresponds to the n^{th} perturbation. A_0 is the initial field; A_1 is the induced field; A_2 is the first order skin effect; and so on.

$$\begin{aligned} A_0 &= (2\pi)^{-1} r^{-1/2} f(x_0) \\ A_1 &= -(2\pi)^{-1} \alpha \int_0^b \int_{t-\ell}^{t+\ell} (r_1/r)^{1/2} [\partial A_{1,0} / \partial z_1] f(x_1) dz_1 dr_1 \\ A_2 &= -(2\pi)^{-1} \alpha \int_0^b \int_{t-\ell}^{t+\ell} (r_2/r)^{1/2} [(\partial A_{2,1} / \partial z_2) + (\partial A_{2,1} / \partial t)] f(x_2) dz_2 dr_2 \end{aligned} \quad \begin{matrix} \text{(Space)} \\ (25s) \end{matrix}$$

A Fortran program for the IEM 7090 has been written to evaluate A_n $(2\pi/\alpha)^n$ ($n = 1, 2, 3$) at $r = 1, z = 0$. Slug length and radius are variable inputs for this program. Typical results are shown in Figure 4. As long as $\alpha < 2\pi$ the approximation

$$A = A_0 + A_1 + A_2 + A_3 \quad (\text{Space}) \quad (26s)$$

seems valid.

3. CONCLUSIONS

The solution as presented in Eq. (25) allows one to calculate the perturbed field around a cylindrical slug moving in a magnetic field. While Eq. (25) is based on the assumption of the A_0 given there, it is easily extended to certain other cases. Any other initial field which has azimuthal independence and has no azimuthal component may be accommodated by suitably changing A_0 . In practice this would be a solenoid or some approximation thereof. If the initial field, hence A_0 is time dependent, a time derivative must be included in the equation for A_1 .

$$A_1 = (2\pi)^{-1} \alpha \int_0^b \int_{t-\ell}^{t+\ell} (r_1/r)^{1/2} [(\partial A_{1,0}/\partial z_1) + (\partial A_{1,0}/\partial t)] f(x_1) dz_1 dr_1. \quad (\text{Space}) \quad (27s)$$

It is possible now to look back and study the difference between the Space Case and the Time Case. The difference first occurs in Eq. (1) where a $\underline{v} \times \underline{B}$ term is added to the electric field in the Space Case. While it appears that one could easily modify the electric field to accommodate this term, in Eq. (8) this term gives rise to a z derivative of the potential. In Eq. (14) the difference in these two cases by the inclusion of the variable t in the limits of the integral is the Space Case. Comparing the Space solution in Figure 4 with the Time solution in Eq. (27t) shows that two cases, which were apparently analogous in Eq. (1), actually have little or nothing in common. The Time Case solution is the sum of a number of sinusoidal waves of phase $0, \pm\pi/2$, and π . The Space Case solution is the sum of some definitely nonsinusoidal functions of no simple phase relation.

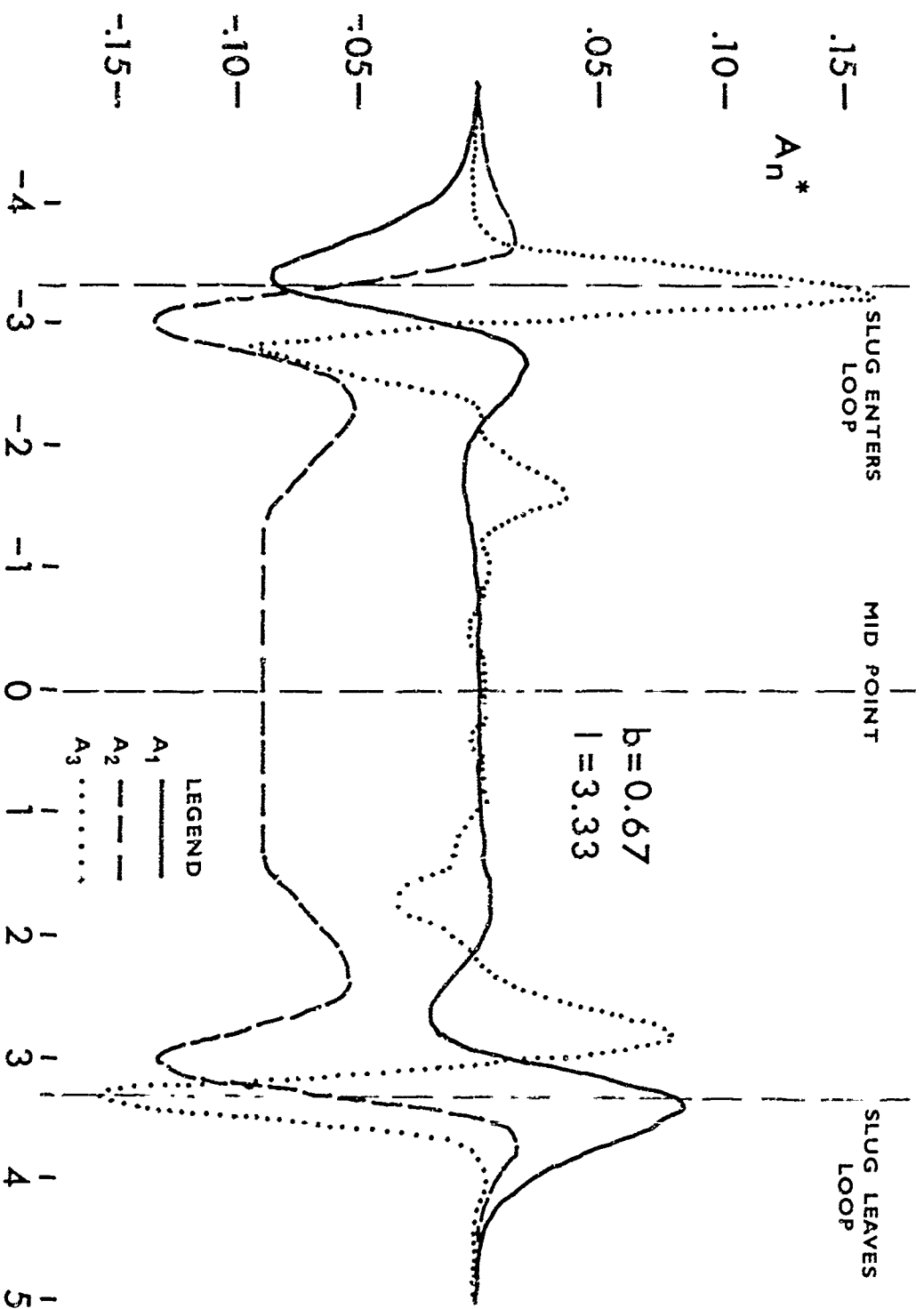


Figure 4. $A_1 (2\pi/\alpha)$, $A_2 (2\pi/\alpha)^2$, and $A_3 (2\pi/\alpha)^3$ as Functions of Slug Position for a Slug of Radius $b = 0.67$, and Length $l = 3.33$

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Appendix A

Singularities

Before a computer can be used to solve Eq. (25), the singular points must be considered. Otherwise the computer may try to evaluate infinity. In Eq. (12), it will be noted that a_n depends on $(r x_n)^{-1/2}$ and K_n . The first of these grows without bound when $r x_n \rightarrow 0$, while K_n grows without bound when $x_n \rightarrow 1$. The first of these is easily handled. Near $r = 0$, $x_n \propto r$. Thus $(r x_n)^{1/2} \propto x_n$. An expansion of the functions in powers of x_n near $x_n = 0$, shows that

$$(2 - x_n) K_n - 2 E_n \propto x_n^2. \quad (A1)$$

Thus, at these points a_n approaches zero. Although this is obvious from physical considerations, the computer must be given mathematical reasons. When $x_n = 1$, a true singularity is found.

$$K_n \propto \ln [4 (1 - x_n^{1/2})^{-1}]. \quad (A2)$$

This is the point at which $r = r_n$, $z = z_n$. In Eq. (10) the current cross section was assumed infinitesimal. This introduces a singularity into the current density. In the case of A_0 one could introduce the wire cross section but the easiest solution is to avoid the singular point. The slug will never run through the current-carrying element so the true value of A_0 at that point is unimportant. In the other a_n this

A2

singularity occurs in a function which is to be integrated. Since K_n is proportional to the logarithm, it is to be expected that the integral of K_n will remain finite. Physically this is expected to be true since infinite fields are not found in the slug. It appears to be true mathematically. While the integrals of Eq. (25s) have not been solved analytically, the following integrals can be solved and are all found to be finite:

$$\begin{aligned} \int_0^1 K_n dx_n \\ \int_0^1 K_n x_n dx_n \\ \int_0^1 K_n x_n^2 dx_n. \end{aligned} \tag{A3}$$

In each case the approximation

$$\int_0^1 K_n g(x_n) dx_n = \int_0^{1-\epsilon} K_n g(x_n) dx_n + \epsilon [K_n g(x_n)]_{x=1-\epsilon} \tag{A4}$$

gives excellent results. With $\epsilon \leq 10^{-2}$ the error is less than 1 percent.

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